

Design & Simulation Of Om Shape Antenna For 4g Applications

Research Scholar Sonia, Asst.Prof. Gaurav Nagpal

Department of Electronics & Communication
MRKIET , Rewari

Abstract- This research article reports a new Dielectric Resonator Antenna (DRA) with its Dielectric resonator (DR) modified to an "OM" shape for UWB (3.1-11.1 GHz), to support high data rate multimedia applications for 4G/5G communications. The proposed DRA reports a peak gain of 7.68 dB and a dual polarization behavior for a frequency band from 6 to 11.1 GHz. It has overall antenna dimensions of $50 \times 40 \times 4.87$ mm³ and is fabricated on a commercially available Rogers RT 5880 substrate (with $\epsilon_r = 2.2$), which is fed using a microstrip feedline with a P-type transformer that offers an input impedance of 50Ω to the DR. A conformal strip between the feedline and the OM shaped DR improves the impedance matching at the UWB frequency response of the DRA. The proposed DRA therefore exhibits an elliptically polarized behavior with axial ratio bandwidth of 5.1 GHz (≤ 10 dB) from 6 to 11.1 GHz. A measured impedance bandwidth of 5.25 GHz from 3.8 to 9.05 GHz and 1.5 GHz from 10 to 11.5 GHz and a peak-measured gain of 7.68 dB at 10.5 GHz (with an average gain of 4.6 dB) has been reported for the proposed DRA. An UWB performance, with good gain properties and an elliptically polarized behavior allows the proposed "OM" shaped DRA to be suitable for short range 4G/5G UWB wireless communications for future multimedia rich WPAN (wireless personal area networks), WLAN, Wi-MAX, INSAT applications, satellite applications, and X band RADAR (for defense communication) applications.

Keyword- Om Shape Antenna

I. INTRODUCTION

Resonators are important components in microwave communication systems. They function as frequency determining components in filters, oscillators, amplifiers, tuners etc. A resonator is said to be 'resonating' at a particular frequency when the electromagnetic energy is stored inside the resonator like an inductor-capacitor-resistor (LCR) circuit. As in the case of LCR circuit, here too, the impedance of the resonator is purely real at the resonant frequency. At frequencies < 100 MHz, Quartz resonators are used.

Recent advances in the communication systems have increased the number of transmitter and receiver channels in a particular geographical area. Therefore, one has to choose the microwave frequency range in order to prevent interferences due to the crowding of channels. The global satellite system, now in operation, uses microwaves for the transmission of meteorological data, telemetry and for international telephone and television channels [1]. The useful range of operation is limited by the allocated frequencies and noise. The extra-terrestrial radiation (sky noise) is a minimum near 4 GHz. It is possible to generate microwave frequencies using a quartz resonator by using frequency multipliers.

However, this increases the noise. In the microwave frequency range, cavities made up of copper, brass, invar etc. are used as resonators taking the advantage of metallic properties. The resonant frequency is determined by the cavity dimensions. The progress in integrated circuit technology led to the miniaturization of components used in microwave communication circuits. Since cavity resonators were bulky and could not be integrated, the microstrip resonators were the only solution to the problem of miniaturization of circuits. But microstrip resonators have poor temperature stability of resonance and low quality factor when compared with metallic cavity resonators. Thus miniaturization was possible only by sacrificing quality factor and temperature-stability. But recent progress in electronic ceramics technology has proved that temperature-stable, high Q resonators for almost all MIC applications can be realized. It is fold that ceramic materials known as 'Dielectric Resonators (DRs)' are very useful for application in the microwave frequency range[1 -18].

DRs possess high dielectric constant (ϵ_r), high quality factor (Q) and small temperature coefficient of resonance (t_r) at microwave frequency. Devices employing DRs have low cost, small size, light weight and high efficiency and are easy-to-use. The versatility and adaptability of DRs have made them indispensable components for filters [19-30], oscillators (DRO) [22,31-39], duplexers etc. in modem microwave integrated circuit (MIC) applications.

Hence the recent advances in Cellular radio, Satellite TV, Military radar etc. have raised the demand for DRs. Materials technology, especially electronic ceramics technology, has gained a lot of attention from scientists who are in search of new materials to satisfy the demands of communication explosion. The global annual sales market of dielectric ceramics, including piezoelectric, ferrites, IC packages etc., used by the industries is expected to exceed 8,500 million by the year 2000. It is now possible to have small and compact resonators for applications from 800 MHz to 100 GHz frequency range.

II. ANTENNA GEOMETRY

Figure 1 (a) ,(b) shows the top and back view of the proposed UWB dual polarized "OM" Shaped DRA,

respectively, designed using a 3D EM tool HFSS. The 3D sideview of the proposed DRA is also shown in Figure 1 (c). The proposed UWB DRA is made from an "OM" shaped block of Alumina (Al₂O₃ with $\epsilon_{DR} = 9.8$, height = 4 mm and $\tan \delta = 0.002$) pasted over a Rogers 5880 substrate sheet (with $\epsilon_r = 2.2$, height = 0.8 mm and $\tan \delta = 0.0009$) with dimensions of $L_s \times W_s$ as 50 mm \times 40 mm. Figure 4.1 (a) ,(b) shows the top and back view of the proposed UWB dual polarized "OM" Shaped DRA, respectively, designed using a 3D EM tool HFSS.

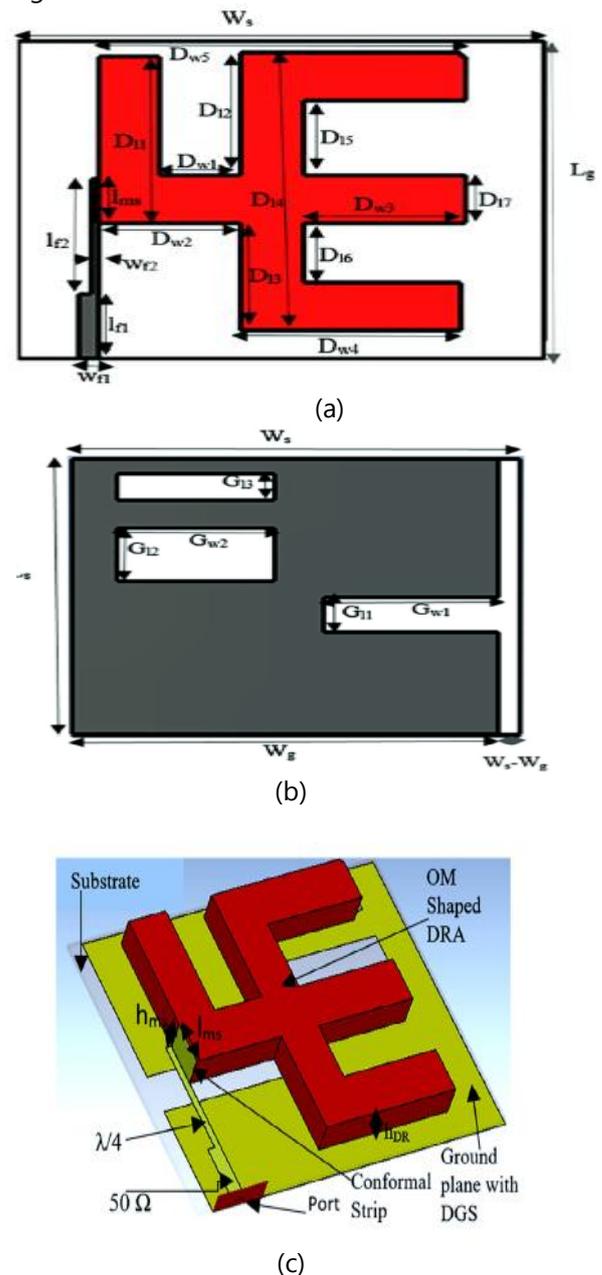


Fig 1 "OM" shaped DRA. (a) Top view. (b) Back view. (c) 3D view of the proposed 'OM' shaped DRA

The 3D sideview of the proposed DRA is also shown in Figure 1 (c). The proposed UWB DRA is made from

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It is the "OM" shaped DR with a DGS, that excites the desired UWB from 3.1 to 11.1 GHz by exciting two higher order modes TE₂₁₁ and TE₂₂₁ at 6.5 and 10.5 GHz, respectively, and presents a simulated improved peak gain of 7.66 dB at 11 GHz. This "OM" shaped DR also generates both horizontal and vertical components of 'E' field giving rise to elliptically polarized waves (axial Ratio < 10 dB), for a frequency band from 6 to 11.1 GHz.

The DR in the antenna is fed using a P type transformer based microstrip feedline as shown in Figure 1C with a length ($l_{f1} + l_{f2}$) as 22.5 mm, offering an impedance of 50 Ω at the antenna's drive point. A conformal rectangular metallic strip of copper with length 5.7 mm and height 4 mm with impedance of 50 Ω is connected between the feedline and the DR. This has one of its edge on the feedline and the other edge on the drive point of the DR. This allows better coupling of EM fields to the DR and hence an improved impedance bandwidth is achieved. The bottom of the Rogers substrate has a metallic ground layer of copper with three slots cut out from it that make it a DGS.

This DGS allows the formation of multiple resonant current loops in the ground layer at the same or nearby frequencies of DR's resonance and thus helps in improving the impedance bandwidth of the antenna's operation. The detailed design parameters of the proposed OM shaped DRA are labeled in Figure 1AC and their optimized parametric values are mentioned in Table 4.1. The next section of this article explains the DRA design process and the equations that are used to obtain the optimized DRA parameters.

III. PROPOSED DRA DESIGN IN HFSS

The proposed antenna is designed with om shaped dielectric resonator antenna. There is two layers of antenna. One is substrate and another one is om shape dielectric resonator antenna.

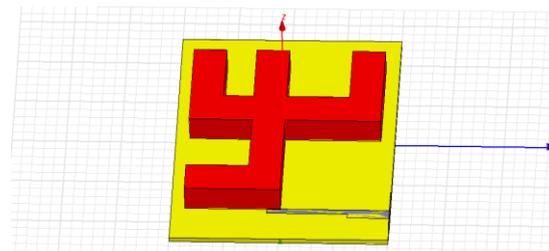


Fig 2 Top View of Proposed Antenna.

The ground layer of proposed antenna is shown in the fig 2. In this layer 3 slots are cut from the ground & these cuts act as defected ground structure.

IV. RESULT ANALYSES

Scattering parameter is one of the parameter which represents proper impedance matching at particular frequency. It is also known as return loss. Return loss may be defined as ratio of reflected back power to the incident power. It is represented in db. In ideal case, return loss must be -10db. As, return loss is minimum, the antenna characterization used to measure.

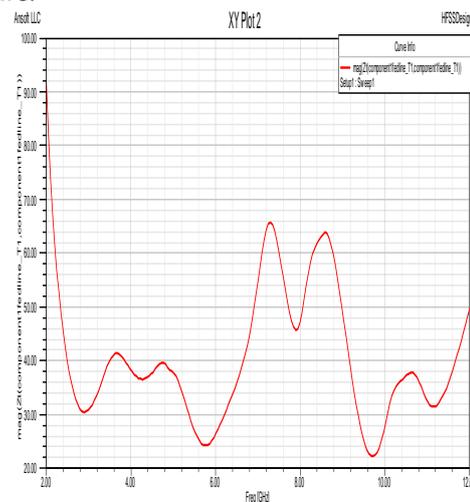


Fig 3 Impedance Matching.

The impedance matching of proposed DRA is shown in fig 3. The impedance bandwidth is analyzed from 2 GHz to 12 GHz. This graph shows that 50 Ω impedance matching is found out at 7.8416 GHz. $0.7604 - j0.224$, $0.7274 + j0.2060$ and $0.7186 - 0.3117 j$

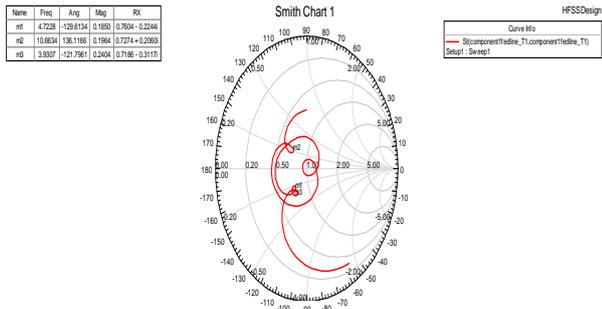


Fig 4 Smith Chart.

The complex graph is represented by Smith chart. The smith chart is analyzed with three frequencies 4.7228, 10.6634 and 3.9307 GHz respectively. The received signal value is $0.7604-j0.224$, $0.7274+j0.2060$ and $0.7186 -j0.3117$ j respectively.

V.CONCLUSION

The design & simulation of different shape of DRA antenna had being studied using HFSS software. The simulation work gives good result through patch calculator. Here accurate patching via resonant frequency has been done. The investigation has been completely done throughout simulation work. The operating frequency is 8 GHz. The proposed Om Shap/e DRA has high gain & best resonant frequency i.e. 7.84 GHz as compared to other shapes of DRA antennas. The proposed Patch Antenna with Ground Slot Antenna have minimum RL i.e -27.7095 db at its resonant frequency 7.84 GHz . Bandwidth of proposed antenna is also improved up to 5.7-11.1 GHz.

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